

Session 19

Outline

- Extrapolation of the autocorrelation ~~of~~ for a SoS signal
- Basics of Autoregressive (AR) spectral estimation
(most widely used parametric spectral estimation technique)
- Problem with non-parametric spectral estimators:

inherently assume $r_{xx}[m] = 0$ for $m > M \leftarrow \text{Block length}$
where $M \leq N-1$

- In fact, for SoS (w/o noise)

$$r_{xx}[m] = \sum_{i=1}^P P_i e^{j\omega_i m}$$

does NOT decay to zero as $m \rightarrow \infty$.

- in general, sharp spectral peaks in $S_{xx}(\omega)$ (true spectrum) imply $r_{xx}[m]$ takes a "long time" to decay to zero as $m \rightarrow \infty$.

- interesting observation re: $r_{xx}[m]$ for SoS model:

$\Rightarrow$ it satisfies recursive relationship:

$$r_{xx}[m] = - \sum_{k=1}^P a_k r_{xx}[m-k] \quad \text{for } m \geq P$$

{ this means, if there are P sinewaves, the true autocorrelation value of lag value m is a linear combination of $r_{xx}[m-1]$, $r_{xx}[m-2]$, ..., all the way to P , $r_{xx}[m-P]$. }

- Where a_k , $k=1, 2, \dots, P$, are the coefficients of the polynomial $(z - e^{j\omega_1})(z - e^{j\omega_2}) \dots (z - e^{j\omega_P})$

$$= z^P + a_1 z^{P-1} + \dots + a_{P-1} z + a_P$$

$\Rightarrow 1 + a_1 z^{-1} + a_2 z^{-2} + \dots + a_{P-1} z^{-(P-1)} + a_P z^{-P} = 0$ for $z_i = e^{j\omega_i}$ ↙ a root of the polynomial

$$1 + a_1 e^{j\omega_i(-1)} + a_2 e^{j\omega_i(-2)} + \dots + a_{p-1} e^{j\omega_i(-(p-1))} + a_p e^{j\omega_i(-p)} = 0.$$

• multiply both sides by $p_i e^{j\omega_i m}$ ($i=1, 2, \dots, p$)

• Sum over all i (since $\sum 0 = 0$)

$$\begin{aligned} \sum_{i=1}^p p_i e^{j\omega_i m} + a_1 \sum_{i=1}^p p_i e^{j\omega_i (m-1)} + \dots + a_{p-1} \sum_{i=1}^p p_i e^{j\omega_i (m-(p-1))} \\ + a_p \sum_{i=1}^p p_i e^{j\omega_i (m-p)} = 0. \end{aligned}$$

Without noise, $\sum_{i=1}^p p_i e^{j\omega_i m} = r_{xx}[m]$

$$\sum_{i=1}^p p_i e^{j\omega_i (m-1)} = r_{xx}[m-1]$$

$$\vdots$$

$$\sum_{i=1}^p p_i e^{j\omega_i (m-p)} = r_{xx}[m-p]$$

• ultimately,

$$r_{xx}[m] = - \sum_{k=1}^p a_k r_{xx}[m-k] \quad \text{for } m \geq p.$$

• QED.

• Thus, if:

- if we estimate $r_{xx}[m]$ for $m=0, 1, \dots, P$ AND
- determine the coefficients a_k , $k=1, 2, \dots, P$.

• We could extrapolate $r_{xx}[m]$ for all m , thereby averting a windowing effect

• ostensibly use recursion to determine $r_{xx}[m]$ for $m \geq P$ and form
$$S_{xx}(\omega) = \sum_{m=-\infty}^{\infty} r_{xx}[m] e^{-j\omega m}$$

• However, in case of SOS, typically interested in frequencies, ω_i , which may be determined from roots of $z^P + a_1 z^{P-1} + \dots + a_{P-1} z + a_P$.

- Note: $r_{xx}[-m] = r_{xx}^*[m]$

- How do we estimate a_k , $k=1, 2, \dots, P$.

- $r_{xx}[m] = -\sum_{k=1}^P a_k r_{xx}[m-k]$

- Write out P equations with P unknowns $\Rightarrow m=1, 2, \dots, P$.

$$\begin{bmatrix} r_{xx}[0] & r_{xx}[-1] & \dots & r_{xx}[-(P-1)] \\ r_{xx}[1] & r_{xx}[0] & \dots & r_{xx}[-(P-2)] \\ \vdots & \vdots & \ddots & \vdots \\ r_{xx}[P-1] & r_{xx}[P-2] & \dots & r_{xx}[0] \end{bmatrix} \begin{bmatrix} a_1 \\ a_2 \\ \vdots \\ a_P \end{bmatrix} = - \begin{bmatrix} r_{xx}[1] \\ r_{xx}[2] \\ \vdots \\ r_{xx}[P] \end{bmatrix}$$

$$\Rightarrow \underset{\substack{\uparrow \\ P \times P}}{R} \underset{\substack{\uparrow \\ P \times 1}}{a} = - \underset{\substack{\uparrow \\ P \times 1}}{r} \quad \left(\text{See SoSextrap.m} \right)$$

- these observations for the SoS model are the basis for the following spectral estimation techniques:

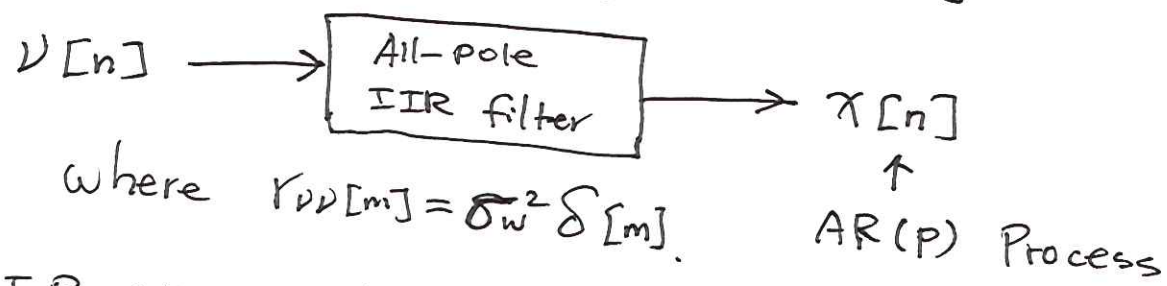
- AutoRegressive Method (AR)

- MUSIC (Multiple Signal Classification) sect. 12.5.3

- ESPRIT (Estimation of Signal Parameters by Rotational Invariance Techniques)
Sect. 12.5.4

• AR Spectral Estimation

- for sole purpose of matching the spectral density of a WSS random process, $x[n]$, we model $x[n]$ as having been generated as:



• IIR means it's a difference equation

- $$x[n] = - \sum_{k=1}^P a_k x[n-k] + v[n]$$
- Can show that $r_{xx}[m]$ satisfies:

$$r_{xx}[m] = - \sum_{k=1}^P a_k r_{xx}[m-k] + \sigma_w^2 \delta[m]$$

- very similar to what we have in SoS.
- A signal generated by passing a white noise through an all-pole ~~IIR~~ filter. Such a signal satisfies the recursive relation.

- in frequency domain:

$$X(\omega) = H(\omega) V(\omega)$$

- where

$$H(\omega) = \frac{1}{1 + \sum_{k=1}^P a_k e^{-jk\omega}}$$

- from basis random process theory

$$S_{xx}(\omega) = |H(\omega)|^2 S_{vv}(\omega)$$

where

$$\begin{aligned} S_{vv}(\omega) &= \text{DTFT} \{ r_{vv}[m] \} \\ &= \text{DTFT} \{ \sigma_w^2 \delta[m] \} = \sigma_w^2 \quad \forall \omega \end{aligned}$$

- thus,

$$S_{xx}(\omega) = \text{DTFT} \{ r_{xx}[m] \}$$

$$= \frac{\sigma_w^2}{\left| 1 + \sum_{k=1}^P a_k e^{-jk\omega} \right|^2}$$