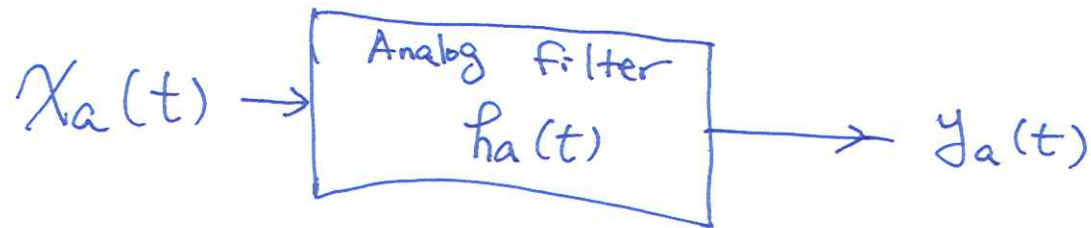


Session 14

Outline

- Characteristics of some common analog filter design Techniques (8.3.5 in P&M)
- Bilinear Transformation Method of Design IIR Filters (Section 8.3.3 in P&M Text)



$$\mathcal{L}\{h_a(t)\} = H_a(s) = \frac{Y_a(s)}{X_a(s)} = \frac{\prod_{k=1}^M (s - z_k)}{\prod_{k=1}^N (s - p_k)}$$

$$H_a(s) = \frac{\sum_{k=0}^M \beta_k s^k}{\sum_{k=1}^N \alpha_k s^k} \quad \alpha_N = 1$$

(polynomial representation)

- $s = \sigma + j\Omega$ Ω - analog angular frequency
 $f_\Omega = \frac{\Omega}{2\pi}$
- Map from s-plane to z-plane ($z = re^{j\omega}$)

$$H(z) = H_a(s) \Big|_{s = \frac{\ln(z)}{d(z)}}$$

- Rational $H_a(s)$ leads to rational $H(z)$ which may be implemented as a difference equation.

- Mapping should also possess the following Properties:

- ① The $j\omega$ axis in s -plane mapped onto $(1 \rightarrow -1)$ unit circle in z -plane

- Guarantees equip-ripple (or monotonic-decreasing) property preserved through transformation

- Note: 1-to-1 mapping between each F in $-\infty < F < +\infty$ and each ω in $-\pi < \omega < \pi$.

② LHP of s -plane mapped into inside of unit circle.

- Guarantees poles in LHP of s -plane mapped into poles inside unit circle in z -plane
- Stable analog filter mapped to stable digital filter

- Bilinear Transformation:

$$S = c \frac{z-1}{z+1} \quad c - \text{constant}$$

this transformation ensures the preservice of the above properties.

- Investigate if mapping satisfies two desired properties:

$$S = \sigma + j\omega \Rightarrow z = re^{j\omega}$$

$$\sigma + j\omega = c \frac{re^{j\omega} - 1}{re^{j\omega} + 1} \left(\frac{re^{-j\omega} + 1}{re^{-j\omega} + 1} \right)$$

$$= c \cdot \frac{(r^2 - 1) + j2r \sin \omega}{r^2 + 2r \cos \omega + 1}$$

- Equating real & imaginary parts on both sides of eqn.:

$$\sigma = c \frac{r^2 - 1}{r^2 + 2r \cos \omega + 1}$$

$$\Omega = c \frac{2r \sin \omega}{r^2 + 2r \cos \omega + 1}$$

- note: $r^2 + 2r \cos \omega + 1 > 0$

- Observe:

$\sigma < 0$ (LHP) dictates:

$$r^2 - 1 < 0 \Rightarrow r^2 < 1 \Rightarrow |r| < 1.$$

- LHP s-plane mapped into inside of unit circle.

- Consider $\sigma = 0 \Rightarrow$ implies ~~that~~ $r = 1$.

$\Rightarrow j\Omega$ -axis in s -plane is mapped to unit circle in z -plane.

$$s = re^{j\omega} = r\cos\omega + jr\sin\omega$$

$$s = \sigma + j\Omega \quad \sigma = 0 \Rightarrow s = j\Omega$$

- Setting $r = 1$ in 2nd expression

$$\Rightarrow \Omega = c \frac{2\sin\omega}{2 + 2\cos\omega} = c \frac{\sin\omega}{1 + \cos\omega}$$

$$\Rightarrow \Omega = c \frac{\sin \frac{\omega}{2} \cos \frac{\omega}{2}}{\cos^2 \frac{\omega}{2}} = c \tan\left(\frac{\omega}{2}\right)$$

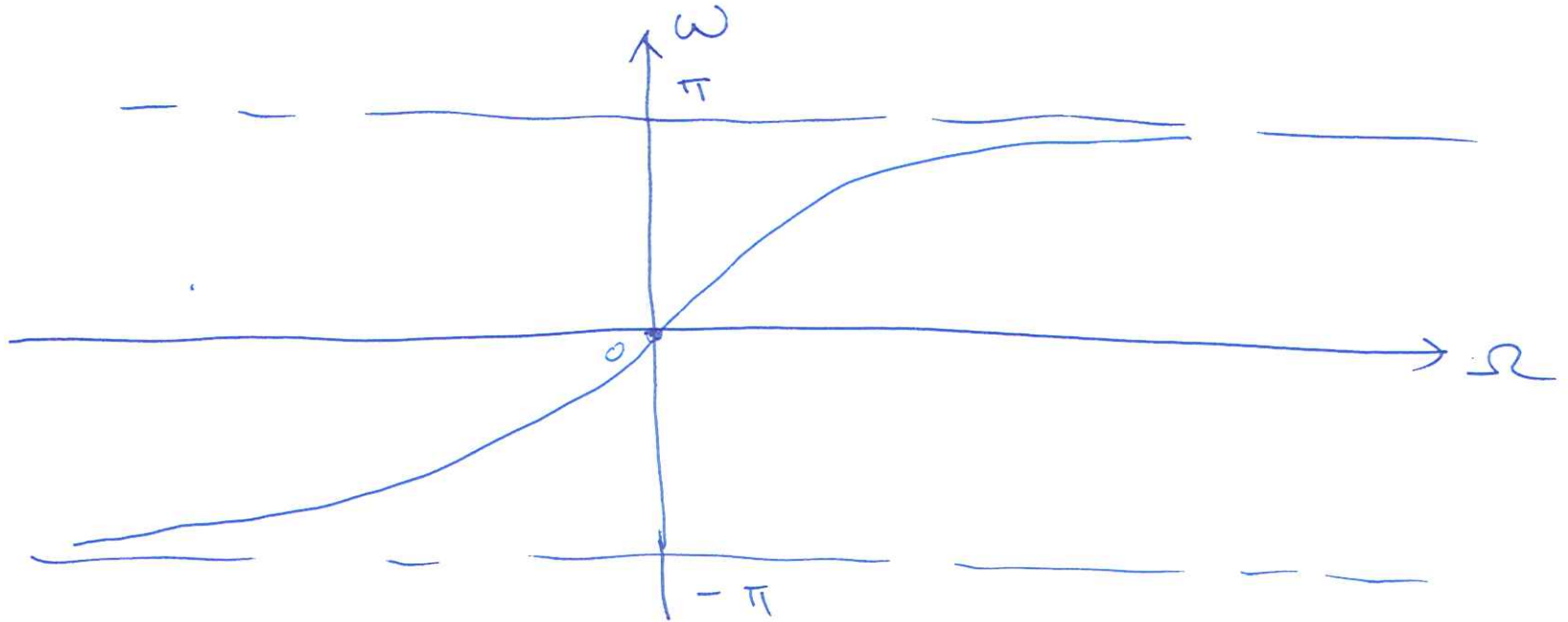
recall: $\cos^2\left(\frac{\omega}{2}\right) = \frac{1}{2}(1 + \cos\omega)$
 $\sin^2\left(\frac{\omega}{2}\right) = \frac{1}{2}(1 - \cos\omega)$

$$\Rightarrow \boxed{\omega = 2 \tan^{-1} \left(\frac{\Omega}{2} \right)}$$

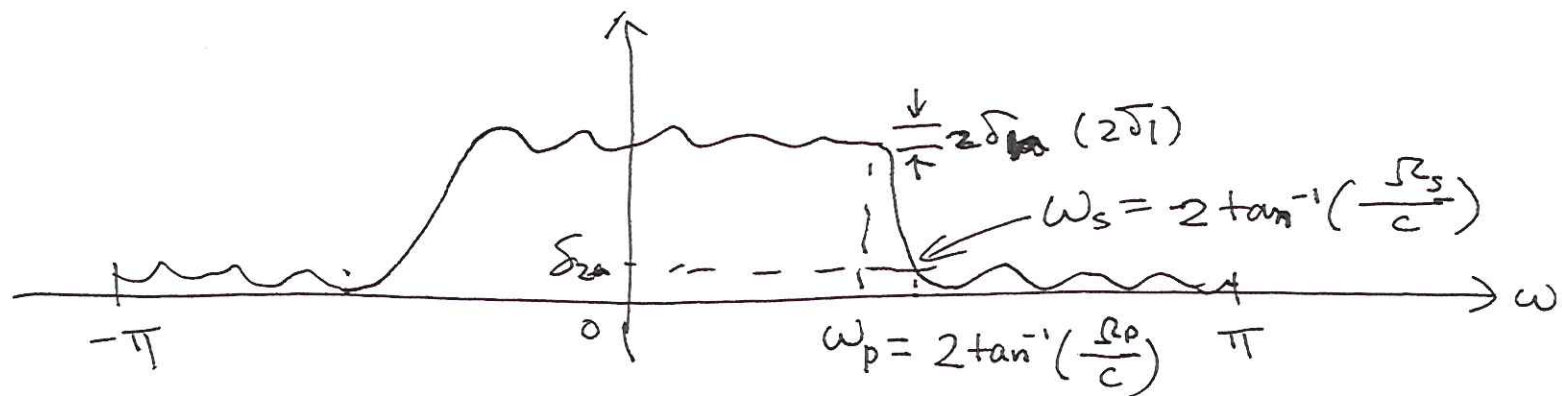
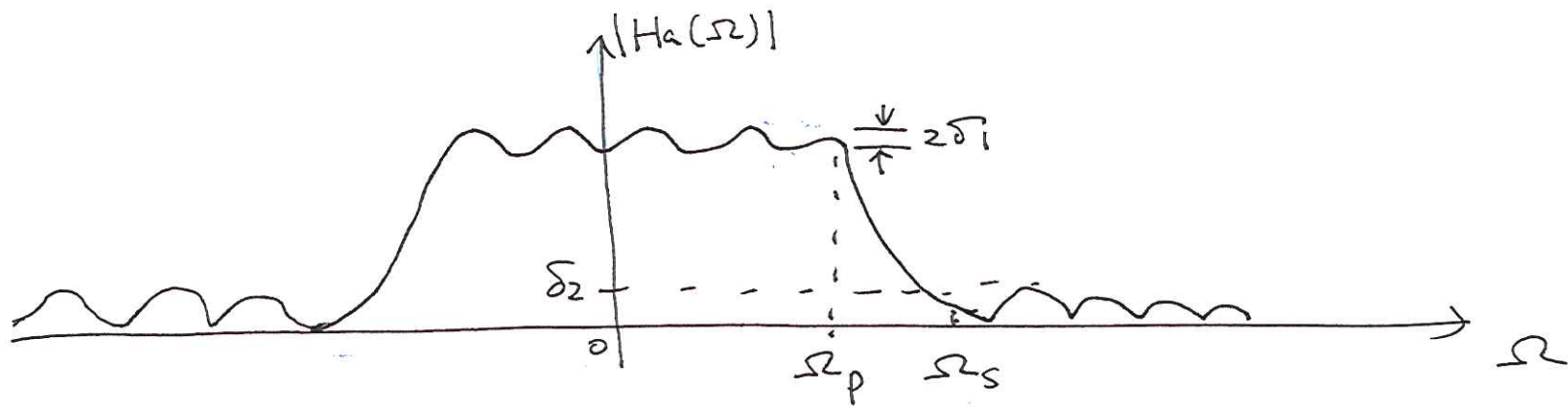
$$\Omega \rightarrow \infty \Rightarrow \omega \rightarrow \pi$$

$$\Omega \rightarrow -\infty \Rightarrow \omega \rightarrow -\pi$$

$$\Omega = 0 \Rightarrow \omega = 0$$



- 1-to-1 mapping as desired
- mapping is non-linear
- frequency compression (warping)



• Design Procedure:

1. Transform design specs in digital domain to analog domain: $\Omega_p = c \tan(\frac{\omega_p}{2})$, $\Omega_s = c \tan(\frac{\omega_s}{2})$
 $\delta_{1a} = \delta_1$, $\delta_{2a} = \delta_2$
2. design analog LPF to meet these specs
 (cookbook design - matlab does it for you).

$$3. \quad H(z) = H_a(s) \Big|_{s=c \frac{z-1}{z+1}}$$

4. Convert to difference equation

- Simple illustrative example:

- design specs: single pole LPF with 3-dB cut-off at $\omega_c = 0.2\pi$.

- 1st order Butterworth filter: $H_a(s) = \Omega_c / (s + \Omega_c)$

$$\Omega_c = c \tan\left(\frac{0.2\pi}{2}\right) = 0.325c$$

$$H(z) = \frac{0.325c}{s + 0.325c} \Big|_{s=c \frac{z-1}{z+1}}$$

$$= \frac{0.325(z+1)}{z-1 + 0.325(z+1)} = \frac{0.245(1+z^{-1})}{1 - 0.509z^{-1}}$$

$$\Rightarrow y[n] = 0.509 y[n-1] + 0.245 x[n] + 0.245 x[n-1]$$